

# Error Mitigation in Noma for Underlay CR Networks with Imperfect Successive Interference Cancellation

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**ABSTRACT-** This study examines the outage probability in a partial relay selection relaying network, an underlay cognitive radio (CR) network, and a non-orthogonal multiple access (NOMA) system. NOMA, which consists of K half-duplex Decode and Forward (DF) relays, is used in the secondary network. These relays are used to enable data transmission to secondary users (SUs) from the secondary base station (SBS). By establishing mathematical formulations, it is feasible to quantify the outage probability that SUs experience while accounting for imperfect successive interference cancellation (i-SIC). The paper also addresses optimum power allocation (OPA) at the SBS, which maximizes the throughput delay in the CR-NOMA system while accounting for the relay of choice independently. The work not only gives mathematical formulas for OPA components but also discusses why these aspects are important for throughput delay improvement. The study highlights how crucial power distribution and relay selection are to enhancing secondary network performance.

**Keywords:** Relay aided NOMA, cognitive radio, Outage Probability, Partial relay selection, imperfect successive interference cancellation, optimal power allocation.

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## 1. INTRODUCTION

The novelty in communication networks is to use a developed CRNOMA scheme, because it can be implemented as a simple add-on. One of the most reliable technologies pertaining to 5G network has taken to be non-orthogonal multiple access (NOMA). It is having the capability for improving spectrum efficiency and overcome the limited user capacity by supporting more users than the available orthogonal resources. [1]. This paper conducts a thorough examination of the cutting-edge CR-based NOMA network architecture, elucidating its fundamental characteristics, merits, and hurdles [2]. High spectral efficiency and large connection requirements need the use of new communication techniques, which are motivated by an exponential rise of mobile devices and the quick expansion of wideband wireless services. It is anticipated that CR and NOMA will play a significant role in fifth generation wireless networks. The potential for increasing system capacity and improving spectral efficiency through an integration of NOMA approaches into CRNs is enormous.

However, employing NOMA causes substantial interference, which presents several technological hurdles. Numerous attempts have been made to make it easier for CRNs to use NOMA and to look at how well [3,4,5] NOMA facilitates the multiplexing of several users over a single resource, with each user being assigned varying power based on their service quality. This study delves into analyzing the throughput of a cognitive radio (CR) system adopting non-orthogonal multiple access (NOMA) techniques, enhanced by the presence of multiple unmanned aerial vehicles (UAVs) serving as relays. The investigation extends to considering both system performance metrics and security constraints [6]. High power coefficients are allocated to weak channel gain and low power coefficients are low power coefficients are allocated to strong channel gain [7]. In NOMA one slot can be used by more than one user so interference will be high as well as so requirement has to remove the interference at the receiver so complexity will be very high. Basically, the biggest challenge is to allocate power for every user in downlink because distance of each user is different from station [8]. Base station has to decide how much power is to be given to user1, user2 etc. Looking at the distance of user from base station will decide the channel gain then allocate the power to each user after that performs the coding. At receiver will combine all coding called superposition coding [9]. A well-established technique for enhancing the capacity and reliability of wireless network is cooperative relaying [10-11]. Cooperative NOMA has been proposed for enhancing reliability of weak user and coverage of the network, it also improves the diversity with throughput of the system [12]. CR-NOMA data in recent studies has been presented in *table 1*.

Table 1: CR-NOMA data in recent studies

Ref.	Authors	Techniques and descriptions	Observation	Year
3	Wang.Xin et al.	A NOMA system with Simultaneous Wireless Information and Power Transfer (SWIPT) for CR network is used. A Dinkelbach algorithm method was used.	The poor channel performance is not considered.	2019
29	Lina Bariah et al.	Developed a precise estimation for the pairwise error probability (PEP) of secondary users utilizing imperfect successive interference cancellation (SIC).	Future work will focus on power allocation, user clustering, green communication	2019
20	K. Chitti, et.al	Explored the fundamental principle of NOMA, the core concept of cognitive radio, the methodology for spectrum sharing, and outlined the benefits and obstacles associated with NOMA and cognitive radio spectrum sharing.	Additional research, implementation, and analysis are required to address the numerous challenges and complexities associated with integrating NOMA with CR.	2020
30	Dinh-Thuan Do et al.	Relay strategy is used in CR-D2D NOMA network	The proposed work is for two users. This can be extended to multiple users	2019
32	T. Balachander	cooperative spectrum sensing (CSS) in Cognitive Radio Network (CRN) using MIMO NOMA technology is implemented also increase the ergodic capacity of secondary user.	Consider only perfect successive interference cancellation at receiver	2021

A number of writers have been examined an effectiveness for CR networks and NOMA systems in the literature. The authors in [8], [9] study MIMO and NOMA techniques whereas [13], [14] studied millimeter-wave NOMA systems. Additionally, relay systems [15]–[16] and cognitive radio (CR) networks [17] were also implemented with NOMA. Power allocation is one of the major concerns in NOMA, Ineffective power distribution has the potential to greatly increase interference, which lowers user-end data recovery. The power allocation applications in NOMA systems are covered in this study. [18]. A writer in this paper derives the minimal quality of system restrictions on Power Allocation Coefficients for each user. They show that there is a basic set of power allocation coefficients that renders the probability non-zero for requirement of minimum QoS rate  $R_0$  [19]. In [20-21] The spatial modulation (SM) that improves spectral efficiency employing a detector for maximum likelihood (ML). It easily manages several CEUs or several CCUs. SM is a modulation method that uses antennas as its indexes to send data. This paper [22] proposed a new type of downlink non-orthogonal multiple access assisted multiple input multiple output system with finite alphabet inputs, where the system's possible spectral efficiency is characterized by mutual information. In order to offer an estimate of the simulated MI, they also suggest a computationally efficient bottom bound. In [23] the authors have shown a method for alternating between spatial modulation with NOMA for same channel gain and power of different users, Results of the simulation indicate that higher spectral efficiency is possible. And authors of [24] studied that Software defined radio (SDR) is the radio communication technology that comes before cognitive radio. The three cognitive radio approaches are Interweave, Overlay, and Underlay. Both the primary non-cognitive radio user and the secondary cognitive radio user can broadcast concurrently

using underlay and overlay techniques. It is not feasible to transmit simultaneously using the interweave approach. In [25-28] authors described nonlinear equations, energy efficiency by optimal fusion rule, iterative search algorithm for maximization of energy efficiency with different design parameters.

## 2. SYSTEM AND CHANNEL MODEL

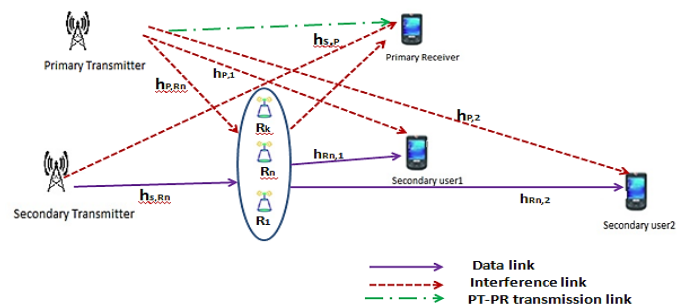


Figure 1. Cognitive Radio Network based on NOMA with primary and secondary transmitters

Here figure1 shows an underlay CR-NOMA system including secondary network, primary network and relays. There are using one of the  $K$  DF HD relays for support.  $\{R_k\}$  where  $k = 1, 2, 3, \dots, K$ . Using the PRS approach, the relay  $R_n$  that facilitates communication between SBS and SUs is chosen. The principal transmitter (PT)-primary receiver (PR) pair makes up primary network, which uses the licensed spectrum [4-6]. As the result, this study has been taken into consideration a two-user scenario. Assuming that the connections linking nodes  $i$  and  $j$  correspond to the fading channel coefficients  $\{h_{i,j}\}$ . It anticipates separate block Rayleigh fading in every connection. As a result,  $E[|h_{i,j}|^2] = \lambda_{ij}$

and a channel power gains  $|h_{i,j}|^2$  exhibit exponential PDFs. It is assumed that the AWGN variance at every receiver of networks is  $\sigma^2$ . An  $R_n$  relay with a highest channel gain to the Secondary base station has been chosen for communication from the SBS to SU's under PRS system under consideration. In representation, this is in papers [29-31]:

$$R_n = \arg \max_{k=1, \dots, K} |h_S, R_k|^2$$

Here K relays are closer as compared to SBS. Since the fading through SBS- $R_k$  connections is assumed to be independent, the exponential random variables  $|h_S, R_k|^2$ . One may obtain the CDF of  $|h_S, R_n|^2$  as [31].

$$F_{|h_S, R_n|^2}(x) = \Pr_{k=1, 2, 3, \dots, K} \max |h_S, R_k|^2 \leq x \quad (1b)$$

$$= \prod_{k=1}^K \Pr(|h_S, R_k|^2 \leq x) = (1 - e^{-x/\lambda_0})^K \quad (1c)$$

$$= 1 - \sum_{i=1}^K (-1)^{i-1} C_i e^{-ix/\lambda_0} \quad (1d)$$

the PDF of  $|h_S, R_n|^2$  is provided by [31]:

$$f_{|h_S, R_n|^2}(x) = \sum_{i=1}^K (-1)^{i-1} C_i \frac{i}{\lambda_0} e^{-ix/\lambda_0}$$

Consequently,  $P_S$  and  $P_R$  will be chosen as [17,31]

$$P_S = \min(P_{S, \max}, \frac{I_{th}}{|h_{S,P}|^2}) \quad (2a)$$

$$P_R = \min(P_{R, \max}, \frac{I_{th}}{|h_{R,P}|^2}) \quad (2b)$$

Let  $s_1(t)$  and  $s_2(t)$  represent the message signals at time t that are meant for  $SU_1$  and  $SU_2$ , respectively. By assigning different power levels to  $SU_1$  and  $SU_2$ , and power allocation coefficients are  $\alpha_1$  and  $\alpha_2$  along with power allotted equal to  $\alpha_1 P_S$  and  $\alpha_2 P_S$ .

The values of  $\alpha_1 < \alpha_2$  and  $\sum_{i=1}^2 \alpha_i = 1$  indicate that the distant user  $SU_2$  at SBS is assigned a greater power.

$$s(t) = \sqrt{\alpha_1 P_S} s_1(t) + \sqrt{\alpha_2 P_S} s_2(t) \quad (3)$$

The signal received is

$$y_{R_n}(t) = h_{S,R_n} \sqrt{\alpha_1 P_S} s_1(t) + |h_{S,R_n}| \sqrt{\alpha_2 P_S} s_2(t) + |h_{P,R_n}| \sqrt{P_P} s_p(t) + w_{R_n}(t) \quad (4)$$

Let equivalent SINRs are  $\Gamma_{R_n, s_2}$ ,  $\Gamma_{R_n, s_1}$  respectively, which are given by

$$\Gamma_{R_n, s_2} = \frac{\alpha_2 P_S |h_{S,R_n}|^2}{\alpha_1 P_S |h_{S,R_n}|^2 + P_P |h_{P,R_n}|^2 + \sigma^2} \quad (5)$$

$$\Gamma_{R_n, s_1} = \frac{\alpha_1 P_S |h_{S,R_n}|^2}{\beta \alpha_2 P_S |h_{S,R_n}|^2 + P_P |h_{P,R_n}|^2 + \sigma^2} \quad (6)$$

The residual interference at  $R_n$  caused by i-SIC is denoted by a term  $\beta \alpha_2 P_S |h_{S,R_n}|^2$  and the i-SIC factor is represented by  $\beta$  ( $0 \leq \beta < 1$ ). The value of  $s_{R_n}(t)$  is provided by

$$s_{R_n}(t) = \sqrt{\alpha_1 P_R} s_1(t) + \sqrt{\alpha_2 P_R} s_2(t) \quad (7)$$

where the power allocation factor at  $R_n$  is denoted by  $\alpha_i$ .  $y_1(t)$  is received signal at  $SU_1$  with  $y_2(t)$  at  $SU_2$  given as

$$y_1(t) = (\sqrt{\alpha_1 P_R} s_1(t) + \sqrt{\alpha_2 P_R} s_2(t)) h_{R_n,1} + h_{P,1} \sqrt{P_P} s_p(t) + w_1(t)$$

$$y_2(t) = (\sqrt{\alpha_1 P_R} s_1(t) + \sqrt{\alpha_2 P_R} s_2(t)) h_{R_n,2} + h_{P,2} \sqrt{P_P} s_p(t) + w_2(t) \quad (9)$$

Now, the distant user  $SU_2$  uses  $y_2(t)$  and the associated SINR to immediately decode its symbol,  $s_2$ .

$$\Gamma_{2, s_2} = \frac{\alpha_2 P_R |h_{R_n,2}|^2}{\alpha_1 P_R |h_{R_n,2}|^2 + P_P |h_{P,2}|^2 + \sigma^2} \quad (10)$$

For near user  $SU_1$ , SINRs are given by

$$\Gamma_{1, s_2} = \frac{\alpha_2 P_R |h_{R_n,1}|^2}{\alpha_1 P_R |h_{R_n,1}|^2 + P_P |h_{P,1}|^2 + \sigma^2} \quad (11)$$

$$\Gamma_{1, s_1} = \frac{\alpha_1 P_R |h_{R_n,1}|^2}{\beta \alpha_2 P_R |h_{R_n,1}|^2 + P_P |h_{P,1}|^2 + \sigma^2} \quad (12)$$

### 3. ANALYSIS OF OUTAGE PROBABILITY

Let  $SU_1$  and  $SU_2$  goal rates be denoted by  $r_1$  and  $r_2$ , respectively, (bits per channel use-bpcu). Attained rates are half since a communication from SBS to  $SU$ 's requires two separate time periods to be completed. For the effective decoding of  $s_1$  and  $s_2$ ,

the target SINRs are therefore provided by  $u_1^{HD} = 2^{2r_1-1}$  and  $u_2^{HD} = 2^{2r_2-1}$

#### 3.1 Explanation of Outage Probability for near User $SU_1$

The OP that  $SU_1$  encountered in the CR-NOMA network under consideration ( $P_{out,1}$ ) is.

$$P_{out,1} = 1 - \Pr\{\Gamma_{R_n, s_2} \geq u_2^{HD}, \Gamma_{R_n, s_1} \geq u_1^{HD}, \Gamma_{1, s_2} \geq u_2^{HD}, \Gamma_{1, s_1} \geq u_1^{HD}\} \quad (13)$$

Putting the SINR expressions (5)-(6) and (11)-(12) in (13).

$$P_{out,1} = 1 - \Pr\left\{ |h_{S,R_n}|^2 \geq \frac{u_2^{HD}(P_P |h_{P,R_n}|^2 + \sigma^2)}{P_S(\alpha_2 - \alpha_1 u_2^{HD})}, |h_{S,R_n}|^2 \geq \frac{u_1^{HD}(P_P |h_{P,R_n}|^2 + \sigma^2)}{P_S(\alpha_1 - \beta \alpha_2 u_1^{HD})}, \right. \\ \left. |h_{R_n,1}|^2 \geq \frac{u_2^{HD}(P_P |h_{P,1}|^2 + \sigma^2)}{P_R(\alpha_2 - \alpha_1 u_2^{HD})}, |h_{R_n,1}|^2 \geq \frac{u_1^{HD}(P_P |h_{P,1}|^2 + \sigma^2)}{P_R(\alpha_1 - \alpha_2 u_1^{HD})} \right\} \quad (15a)$$

$$= 1 - \Pr\left\{ |h_{S,R_n}|^2 \geq \frac{\phi_0}{P_S} (P_P |h_{P,R_n}|^2 + \sigma^2) \times \Pr\{|h_{R_n,1}|^2 \geq \frac{\phi_1}{P_R} (P_P |h_{P,1}|^2 + \sigma^2)\} \right\} \quad (15b)$$

For simplification assume first part is  $A_0$  and second part is  $A_1$ .

$$= 1 - A_0 A_1 \quad (15c)$$

Now finding the values of  $A_0$  and  $A_1$

$$A_0 = \Pr\left\{ |h_{S,R_n}|^2 \geq \frac{\phi_0}{P_{S, \max}} (P_P |h_{P,R_n}|^2 + \sigma^2), P_{S, \max} < \frac{I_{th}}{|h_{S,P}|^2} \right\} + \Pr\left\{ |h_{S,R_n}|^2 \geq \frac{\phi_0}{I_{th}} (P_P |h_{P,R_n}|^2 + \sigma^2), P_{S, \max} > \frac{I_{th}}{|h_{S,P}|^2} \right\} \quad (16)$$

For simplification assume first part is  $A_{00}$  and second part is  $A_{01}$

$$A_{00} = \Pr\left\{ |h_{S,R_n}|^2 \geq \frac{\phi_0}{P_{S, \max}} (P_P |h_{P,R_n}|^2 + \sigma^2) \right\} \times \Pr\left\{ |h_{S,P}|^2 < \frac{I_{th}}{P_{S, \max}} \right\} \quad (17a)$$

Let  $\lambda_{S,P}$  and  $\lambda_{P,R_n}$  are the mean values of  $|h_{S,P}|^2$   $|h_{P,R_n}|^2$  so  $A_{00}$  becomes

$$A_{00} = \sum_{i=1}^K (-1)^{i-1} C_i \int_{y=0}^{\infty} e^{-\frac{i\phi_0(P_P y + \sigma^2)}{P_{S, \max} \lambda_0} |h_{P,R_n}|^2 (y) dy} \times 1 - e^{-\frac{I_{th}}{P_{S, \max} \lambda_{S,P}}} \quad (17b)$$

$$= \left(1 - e^{-\frac{I_{th}}{P_{S,max}\lambda_{S,P}}}\right) \times \sum_{i=1}^K (-1)^{i-1} K_{C_i} \times \int_{y=0}^{\infty} e^{-\frac{i\phi_0(P_p y + \sigma^2)}{P_{S,max}\lambda_0} \frac{1}{\lambda_{P,R_n}}} e^{-\frac{y}{\lambda_{P,R_n}}} dy$$

(17)c

Simplify it, So A<sub>00</sub> is

$$A_{00} = \left(1 - e^{-\frac{I_{th}}{P_{S,max}\lambda_{S,P}}}\right) \times \sum_{i=1}^K (-1)^{i-1} K_{C_i} \times e^{-\frac{i\phi_0\sigma^2}{P_{S,max}\lambda_0}} \left(\frac{P_{S,max}\lambda_0}{i\phi_0 P_p \lambda_{P,R_n} + P_{S,max}\lambda_0}\right)$$

(17)d

$$P_{out,1} = 1 - \left[ \left(1 - e^{-\frac{I_{th}}{P_{S,max}\lambda_{S,P}}}\right) \times \sum_{i=1}^K (-1)^{i-1} K_{C_i} e^{-\frac{i\phi_0\sigma^2}{P_{S,max}\lambda_0}} \left(\frac{P_{S,max}\lambda_0}{i\phi_0 P_p \lambda_{P,R_n} + P_{S,max}\lambda_0}\right) - \sum_{i=1}^K (-1)^{i-1} K_{C_i} \times \frac{I_{th}\lambda_0}{\lambda_{S,P}\lambda_{P,R_n}i\phi_0 P_p} e^{\frac{I_{th}\lambda_0}{i\phi_0 P_p \lambda_{P,R_n}} \left(\frac{i\phi_0\sigma^2}{I_{th}\lambda_0} + \frac{1}{\lambda_{S,P}}\right)} Ei \left( - \left(\frac{i\phi_0\sigma^2}{I_{th}\lambda_0} + \frac{1}{\lambda_{S,P}}\right) \frac{I_{th}}{P_{S,max}} - \left(\frac{i\phi_0\sigma^2}{I_{th}\lambda_0} + \frac{1}{\lambda_{S,P}}\right) \frac{I_{th}\lambda_0}{i\phi_0 P_p \lambda_{P,R_n}} \right) \right]$$

$$\times \left[ \left(1 - e^{-\frac{I_{th}}{P_{R,max}\lambda_{R,P}}}\right) e^{-\frac{\phi_1\sigma^2}{P_{R,max}\lambda_{R,P,1}}} \times \frac{P_{R,max}\lambda_{R,P,1}}{\phi_1 P_p \lambda_{P,1} + P_{R,max}\lambda_{R,P,1}} - \frac{I_{th}\lambda_{R,P,1}}{\lambda_{R,P}\lambda_{P,1}\phi_1 P_p} e^{\frac{I_{th}\lambda_{R,P,1}}{\phi_1 P_p \lambda_{P,1}} \left(\frac{\phi_1\sigma^2}{I_{th}\lambda_{R,P,1}} + \frac{1}{\lambda_{R,P}}\right)} \times Ei \left( - \left(\frac{\phi_1\sigma^2}{I_{th}\lambda_{R,P,1}} + \frac{1}{\lambda_{R,P}}\right) \frac{I_{th}}{P_{R,max}} - \left(\frac{\phi_1\sigma^2}{I_{th}\lambda_{R,P,1}} + \frac{1}{\lambda_{R,P}}\right) \frac{I_{th}\lambda_{R,P,1}}{\phi_1 P_p \lambda_{P,1}} \right) \right]$$

(17)e

Now for finding the A<sub>01</sub> for equation (16), assuming

$$|h_{p,R_n}|^2 = Y, |h_{s,p}|^2 = Z, \text{ putting the value of } |h_{s,R_n}|^2$$

$$A_{01} = \sum_{i=1}^K (-1)^{i-1} K_{C_i} \int_{y=0}^{\infty} \int_{z=\frac{I_{th}}{P_{S,max}}}^{\infty} e^{-\frac{i\phi_0}{I_{th}\lambda_0} z (P_p y + \sigma^2)} f|_{h_{s,p}}|^2(z) f|_{h_{p,R_n}}|^2(y) dz dy$$

(18a)

$$\sum_{i=1}^K (-1)^{i-1} K_{C_i} \frac{1}{\lambda_{P,R_n}\lambda_{S,P}} \int_{y=0}^{\infty} \int_{z=\frac{I_{th}}{P_{S,max}}}^{\infty} e^{-\frac{i\phi_0}{I_{th}\lambda_0} z (P_p y + \sigma^2)} e^{-\frac{z}{\lambda_{S,P}}} e^{-\left(\frac{y}{\lambda_{P,R_n}}\right)} dz dy$$

(18b)

$$A_{01} = \sum_{i=1}^K (-1)^{i-1} K_{C_i} \frac{I_{th}\lambda_0}{i\phi_0\lambda_{S,P}P_p\lambda_{P,R_n}} \times e^{\frac{I_{th}\lambda_0}{i\phi_0 P_p \lambda_{P,R_n}} \left(\frac{i\phi_0\sigma^2}{I_{th}\lambda_0} + \frac{1}{\lambda_{S,P}}\right)} \times Ei \left( - \left(\frac{i\phi_0\sigma^2}{I_{th}\lambda_0} + \frac{1}{\lambda_{S,P}}\right) \frac{I_{th}}{P_{S,max}} - \left(\frac{i\phi_0\sigma^2}{I_{th}\lambda_0} + \frac{1}{\lambda_{S,P}}\right) \frac{I_{th}\lambda_0}{i\phi_0 P_p \lambda_{P,R_n}} \right)$$

(18c)

From equation (15)c, we can find out A<sub>1</sub> with the help of equation (2)

$$A_1 = Pr\left\{ |h_{R_n,1}|^2 \geq \frac{\phi_1}{P_{R,max}} (P_p |h_{p,1}|^2 + \sigma^2) \right\} \times P_{R,max} < \frac{I_{th}}{|h_{R_n,P}|^2} + Pr\left\{ |h_{R_n,1}|^2 \geq \frac{\phi_1}{I_{th}} (P_p |h_{p,1}|^2 + \sigma^2) \right\} |h_{R_n,P}|^2 P_{R,max} > \frac{I_{th}}{|h_{R_n,P}|^2}$$

(19)

Here first term assumes A<sub>10</sub> and second term is A<sub>11</sub>. As it finds A<sub>00</sub> and A<sub>01</sub> similarly A<sub>10</sub> and A<sub>11</sub> can be find out

$$A_{10} = \left(1 - e^{-\frac{I_{th}}{P_{R,max}\lambda_{R,P}}}\right) \times e^{-\frac{\phi_1\sigma^2}{P_{R,max}\lambda_{R,P,1}}} \times \left(\frac{P_{R,max}\lambda_{R,P,1}}{\phi_1 P_p \lambda_{P,1} + P_{R,max}\lambda_{R,P,1}}\right)$$

(20)

$$A_{11} = - \frac{I_{th}\lambda_{R,P,1}}{\phi_1\lambda_{R,P}P_p\lambda_{P,1}} \times e^{\frac{I_{th}\lambda_{R,P,1}}{\phi_1 P_p \lambda_{P,1}} \left(\frac{\phi_1\sigma^2}{I_{th}\lambda_{R,P,1}} + \frac{1}{\lambda_{R,P}}\right)} \times Ei \left( - \left(\frac{\phi_1\sigma^2}{I_{th}\lambda_{R,P,1}} + \frac{1}{\lambda_{R,P}}\right) \frac{I_{th}}{P_{R,max}} - \left(\frac{\phi_1\sigma^2}{I_{th}\lambda_{R,P,1}} + \frac{1}{\lambda_{R,P}}\right) \frac{I_{th}\lambda_{R,P,1}}{\phi_1 P_p \lambda_{P,1}} \right)$$

(21)

In (15)c, substituting (17)d, (18)c, (20), and (21) yields (14). It is evident from (15)a that 0 < P<sub>out,1</sub> < 1 if and only if

$$0 < \alpha_1 < \frac{1}{1+u_2^{HD}} \frac{\beta u_1^{HD}}{1+\beta u_1^{HD}} < \alpha_1 < 1, 0 < \alpha_1 < \frac{1}{1+u_2^{HD}} \frac{\zeta u_1^{HD}}{1+\zeta u_1^{HD}} < \alpha_1 < 1, \text{ or else } P_{out,1} \text{ can be unity.}$$

If any of these requirements are not met, the probability component on the right-hand side of equation (15)b becomes 0.

### 3.2 Explanation of Outage Probability for far User SU<sub>2</sub>

Similarly, Outage Probability of SU<sub>2</sub>

$$P_{out,2} = 1 - A_0 B_0 = (1 - A_{00} + A_{01}) (B_{00} + B_{01})$$

Now P<sub>out,2</sub> can be find with the help of all equation. Here 0 < P<sub>out,2</sub> < 1 if and only if

$$0 < \alpha_1 < \frac{1}{1+u_2^{HD}} \frac{\beta u_1^{HD}}{1+\beta u_1^{HD}} < \alpha_1 < 1, 0 < \alpha_1 < \frac{1}{1+u_2^{HD}} \text{ or else } P_{out,2} \text{ can be unity. If these conditions are not satisfied then probability can be unity.}$$

## 4. PERFORMANCE EVALUATION RESULTS

The analytical and simulation findings for throughput and OP are presented in this section. The distances between the relays and the SBS are significantly greater than the distances between the relays, hence considering that relays are grouped somewhere in between SUs and SBS. The OP that SU<sub>1</sub> and SU<sub>2</sub> in the CR-NOMA network encounter against P<sub>max</sub> and their respective throughput against transmit power are displayed in figures 2,3.

Firstly, noted that SU<sub>2</sub> experiences less outage probability than SU<sub>1</sub>, as its power allocation factor is less at the SBS and a chosen R<sub>n</sub> relay. Additionally, SUs' OP is lower under the PRS scheme than under the RRS model. When the PRS scheme is used instead of the RRS scheme, An Outage Probability for SU<sub>1</sub> and SU<sub>2</sub> fall near 95% and 89%, correspondingly, with K = 3 and P<sub>max</sub> = 40dB. Additionally, under the PRS system, a greater value of K considerably enhances the OP performance of both SU<sub>1</sub> and SU<sub>2</sub>.

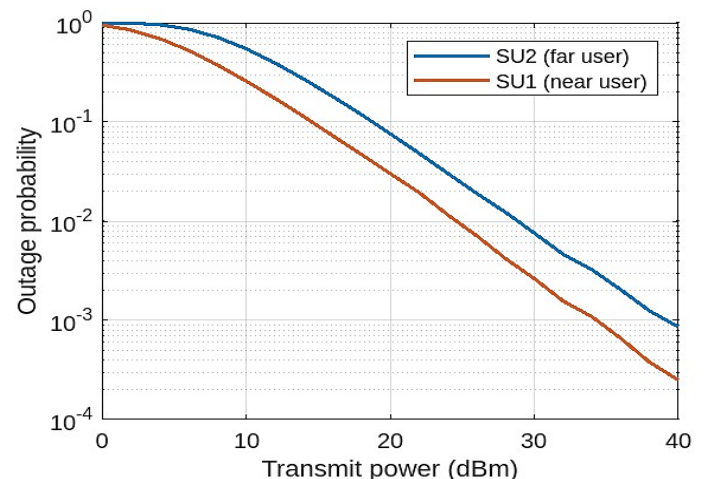
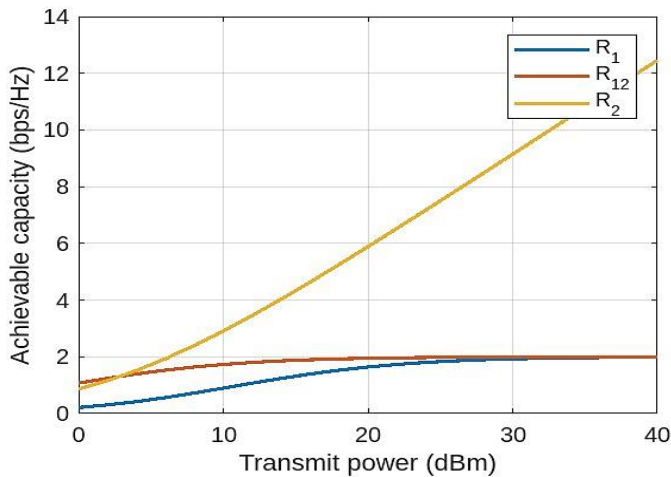
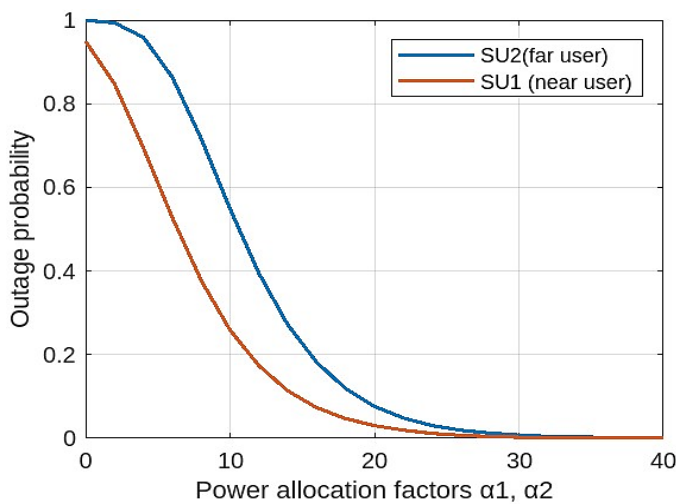


Figure 2. Outage Probability of SU1 and SU2 vs. transmit power




**Figure 3.** capacity of  $SU_1$  and  $SU_2$  vs. transmit power

**Figure 4.** Outage probability of  $SU_1$  and  $SU_2$  vs. Power allocation factor

The secondary network's throughput results in the CR-NOMA system are displayed in *figure 4*. An OP for  $SU$ 's is observed to decrease as  $P_{max}$  is increased, provided that  $P_{max}$  is less than  $I_{th}$ . Result comparison with recent research has been shown in *table 2*.

**Table 2. Result comparison with recent research**

System model / method	Parameters	Maximum EE (bits/Hz/Joule)	Remarks
Wireless powered downlink and multiuser information uplink, Lagrangian multiplier approach [27]	Transmit power	2:55	Maximum EE is achieved for throughput $R_0$ , 0.5 bits/s/Hz
Massive MIMO for hybrid architectures based on phase shifters	Number of base station antenna	0:054	Hybrid architecture of massive MIMO, Maximum EE is

[13]	and SNR		achieved at 300 BS and 20 dB SNR
EE maximization of Cooperative Cognitive radio in fading and non-fading environment [26]	Number of Secondary users, SNR	4:92	Maximum EE is achieved for koN rule in non-fading at throughput $R_0$ , 0.5 bits/s/Hz, SNR — 4 dB
Proposed: outage probability in a partial relay selection relaying network	secondary network, primary network and relays	4:97	Maximum EE is achieved with an Outage Probability for $SU_1$ and $SU_2$ fall near 95% and 89%, correspondingly, with $K = 3$ and $P_{max} = 40dB$

## 5. CONCLUSION

In this work, Paper provides equations for the system throughput and outage probability of an underlay CR-NOMA network using the Partial Relay Selection (PRS) design. The  $SU$ 's experiences are considered. These formulations account for a number of important elements, such as it is planned to broaden the scope of our analysis in the future by taking into account more elements and variables, which will help us better comprehend the functionality and performance of the system.

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